

Offline Primary Side Regulation LED Flyback Controller with PWM Dimming

Features

- Primary Side Regulation (PSR); No Opto-isolator Required
- Primary Side Supply Chip Power, Simplified Transformer Design
- Universal Input Voltage Range (85V 265V)
- Constant Current Output (+/- 3%)
- Excellent PWM Dimming
- High Line & Load Regulation (+/-2%)
- High Efficiency (>85%)
- Open & Short LED Protection
- Primary Side OCP
- VCC Over Voltage Protection (OVP)
- VCC Under Voltage Lockout (UVLO)
- ZCD Short Protection
- CS Sense Resistor Open Protection
- Over Temperature Protection (OTP)
- Cycle by Cycle Current Limit
- Patent Pending Sense Architecture
- SOP-8 Package Available

Applications

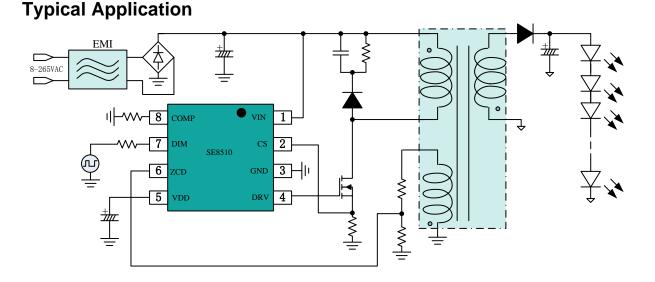
- Isolated LED driver with PWM Dimming
- LED Lighting

Description

SE8510 is an offline isolated flyback led controller with primary side control IC. Boundary mode operation provides a small magnetic solution with excellent load regulation. Using a patent pending current sensing scheme, the controller is able to deliver a well regulated current to the secondary side without using an opto-coupler.

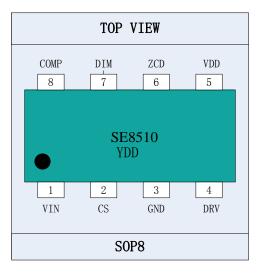
SE8510 uses a rugged high voltage junction isolated process that can withstand an input voltage surge of up to 450V. Then SE8510 can be powered directly by main line input voltage source, this simplify the design of transformer for wide output voltage. Also the total application design becomes simpler than before.

SE8510 has PWM dimming control feature. Additional features include thermal shutdown, current limit, open & short LED load and VDD under-voltage lockout.



Typical SE8510 LED Driver Application Circuit

Pin Configuration



Pin Descriptions

Name	SOP8	Description	
VIN	1	Input voltage 8V to 400V DC	
CS	2	Senses primary current	
GND	3	Device ground	
DRV	4	Drives the gate of the external MOSFET	
VDD	5	Internally regulated supply voltage	
ZCD	6	feedback terminal to detect zero current	
DIM	7	PWM dimming pin	
COMP	8	power compensation	

Order Information

Туре	Package	Logo	Temp	Tape and Reel
SE8510	SOP-8	SE8510 YDD	-40℃ to 85℃	2,500

SE8510, SE- Company Logo, 8510 - Chip Type, Y - Year, DD - Production Code

Absolute Maximum Ratings

VIN to GND	-0.3V to +450V		
CS, VID, DRV, ZCD, LD, DIM, COMP to GND	-0.3V to (VDD+0.3V)		
Junction Temperature	+150°C		
Storage Temperature Range	-65°C to +150°C		

Note: Absolute maximum ratings are limits beyond which damage to the device may occur. Operating Ratings are conditions for

which the device is intended to be functional, but device parameter specifications may not be guaranteed.

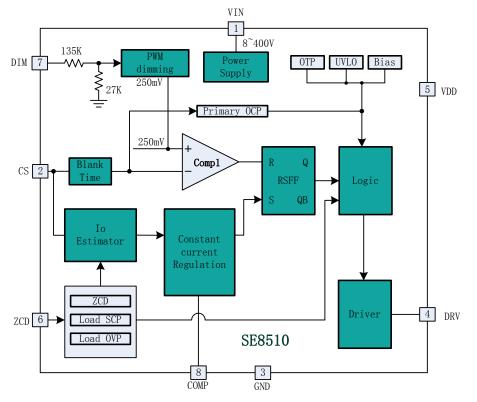
Electrical Characteristics

(TA = 25°C, Vdd = 7.5V, Vin=12V unless noted otherwise)

Symbol	Parameter	Min	Typical	Max	Unit	Condition
V _{IN}	Input DC supply voltage range ¹	8		400	V	DC input voltage
I _{IN}	supply current	-	2	3	mA	
V_{DD}	Internally regulated voltage	7.0	7.5	8.0	V	V _{IN} = 8-400V, pin DRV open
I _{DD}	VDD current for external circuitry	-	-	0.5	mA	V _{IN} = 8 - 100V
UVLO	VDD under voltage lockout threshold	4.5	5.0	6.5	V	V _{IN} rise
ΔUVLO	VDD under voltage lockout hysteresis	-	1.5	-	V	V _{IN} fall
Vcs	Current sense threshold voltage	225	250	275	mV	Temp=-40℃ to +85℃
V _{DRV} (hi)	DRV high output voltage	V _{DD} -0.3	-	V _{DD}	V	I _{OUT} = 5mA
V _{DRV} (lo)	DRV low output voltage	0	-	0.3	V	I _{OUT} = -5mA
R _{DIM}	Pin DIM pull-down resistance		165.0		KΩ	For SE8510
V_{DIM}	Pin DIM voltage range	0	-	5	V	TEMP ≤ 85° C, for SE8510
V _{DIM-th}	Pin DIM threshold voltage range	1.35	1.50	1.65	V	TEMP \leq 85°C, for SE8510
T _{BLANK}	Current sense blanking interval	300	450	600	ns	$V_{CS} \ge 300 \text{mV}, V_{LD} = V_{DD}$
T _{RISE}	DRV rise time	-	30	50	ns	C _{GATE} = 500pF
T _{FALL}	DRV fall time	-	30	50	ns	$C_{GATE} = 500 pF$
V _{ZCD}	ZCD feedback comparator threshold		350		mV	
V _{OVP}	ZCD over voltage threshold		1.2		V	
I _{OCP}	primary side over current threshold		4.8*I _{PK}		А	IPK=primary peak current
T _{SD}	Thermal shutdown temperature		160		°C	
T _{hy}	Thermal shutdown hysteresis		70		°C	

1. Also limited by package power dissipation limit, whichever is lower.

SIMPLIFIED INTERNAL BLOCK DIAGRAM



Detailed Description

SE8510 is an offline isolated flyback LED controller with primary side control IC. It combines a high performance low voltage PWM controller chip and an input high voltage MOSFET that can withstand an input voltage surge of up to 450V. Using a patent pending current sensing scheme, the controller is able to deliver a well regulated current to the secondary side without using an opto-coupler. Moreover it guarantees a safe operation when short circuit of LEDs occurs, or open load circuit.

Supply Power

As shown in block diagram, chip current is internally generated in SE8510 without using bulky startup resistors typically required in the offline applications. Moreover, in many applications the SE8510 can be continuously powered using its internal linear regulator that provides a regulated voltage of 7.5V for all internal circuits.

Its supply is from the main line source. By this way the design of application system becomes simpler than before.

Discontinuous Conduction Mode

Critical conduction mode is a variable frequency switching scheme that always returns the secondary current to zero with every cycle. The SE8510 relies on boundary mode or discontinuous mode to calculate the critical current because the sensing scheme assumes the secondary current returns to zero with every cycle. The ZCD pin uses a patented schematic to determine the zero current point.

Primary-Side Current Control

When the voltage at CS pin exceeds a peak current sense voltage threshold, the gate drive signal terminates, and the power MOSFET turns off. The threshold is internally set to 250mV. The primary peak current can be calculated as:

$$I_{pk1} = \frac{250 \text{mV}}{R_{cs}}$$
....(1)

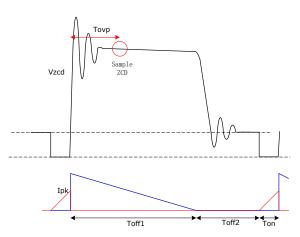


Figure 1. Discontinuous Conduction Mode

Secondary-Side(load) Control



SE8510 controls the secondary side LED current from the primary side information. The output LED mean current can be calculated approximately as:

$$I_{out} = \frac{62.5 \text{m} \times \text{N}}{\text{R}_{cs}}.....(2)$$

or,

$$I_{out} = \frac{I_{pk1} \times N}{4}....(3)$$

N — Turn ratio of primary side to secondary side

 R_{CS} — primary side current sense resistor I_{PK1} — primary side peak current

Switching Frequency

The Switching Frequency of SE8510 can be calculate as:

$$F_{osc} = \frac{V_{out}}{{}_{2NL_{sec}}I_{pk1}}.....(4)$$

or,

$$F_{osc} = \frac{V_{out}}{{}^{8L_{sec}I_{out}}}.....(5)$$

V_{OUT} — output load voltage

 $\rm N-$ Turn ratio of primary side to secondary side

L_{SEC} — secondary side inductance

I_{PK1} — primary side peak current

It is better to set the frequency lower than 70KHz. The recommend value is 45KHz.

PWM dimming Feature (pin DIM)

PWM dimming can be achieved by driving the DIM pin with a low frequency square wave signal. When the PWM signal is zero, the GATE driver is turned off and when the PWMD signal is high(>1.5V), the GATE driver is enabled. The LED current is proportional to the duty cycle of PWM wave.

Output Over Voltage & Open Load Protection (OVP)

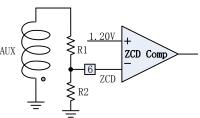


Figure 2. Over Voltage Protection

After the MOSFET turns off, the ZCD comparator detects the level of pin ZCD. Once its level is larger than 1.2v the gate driver will be turned off. To avoid mis-trigger OVP the ZCD OVP function does not work until the blank time Tzcd (fig.1), the typical interval is 1us.

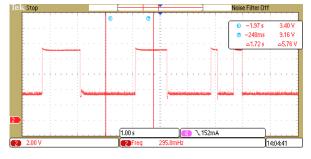
The OVP threshold is:

$$W_{\rm ovp} = 1.2 \times \left(1 + \frac{R_1}{R_2}\right) \frac{N_{\rm sec}}{N_{\rm aux}}$$
.....(6)

N_{SEC}—The secondary winding turns

N_{AUX}—The auxiliary winding turns

The below wave records four OVP events.



Primary-Side Current Limit

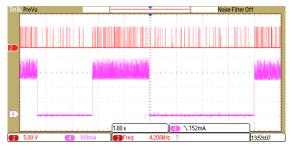
When the primary current exceeds a limit, the chip will shut down, and then wait for a moment to restart. The limit is $4.8*I_{PK}$.

Output short Circuit Protection (OSCP)

Chip can control the output current, but

when OSCP the better way is to disable PWM signal. SE8510 will trigger OSCP after short LED series. After a fixed time the chip will restart to try whether the OSCP is still here. If so, the chip will re-disable PWM signal.

The below wave records two output short events.



Package Information

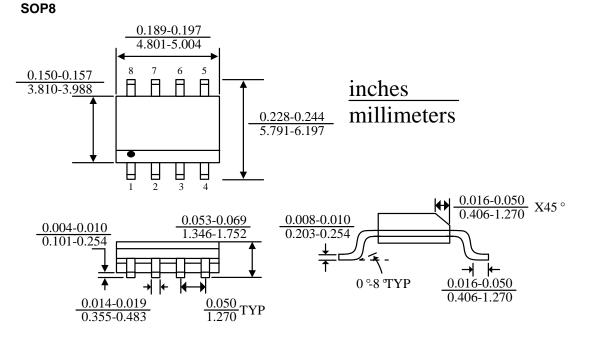
Hysteretic Thermal Shutdown

The thermal shutdown circuitry senses the controller die temperature. The threshold is set at 160 °C typical with a 70 °C hysteresis.

Application Example

Please refer to document AN26 which has one detailed design example, or contact someone of the company.

The final application is shown in figure 3.



Version

Version	Date	Note
V4.0	2013.10.31	Pin DIM spec modified

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Application Schematic

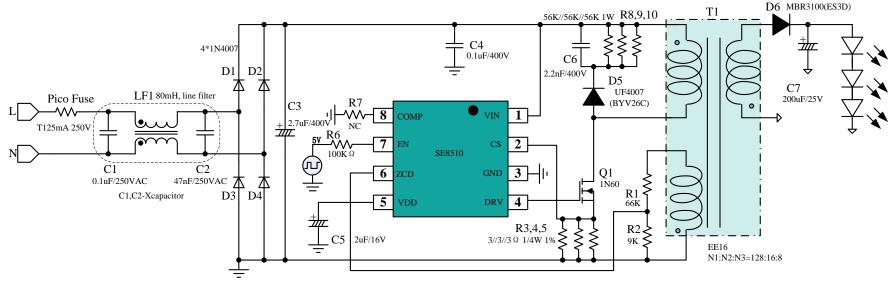


Figure 3. Application Schematic 9.6V/500mA LED driver